IN THE CLAIMS:

Please amend the claims of this application as follows.

1.(currently amended) A method for acoustic, and in particular ultrasonic, receiver

beamforming for use in ultrasonic imaging, said method comprising the steps of:

transmitting by means of at least one electroacoustic transducer at least one beam of

acoustic wave signals into a body under examination, said signals being transmitted at a

first frequency;

receiving said acoustic wave signals reflected by said body under examination

through an array of receiving electroacoustic transducers;

synchronizing said signals received by each of said receiving transducers from one

or more reflection sources arranged in a predetermined area, line or point of said body

under examination by applying delays to said received signals by each of said receiving

transducers, said delays being a function of acoustic wave propagation velocity and of the

geometric distance the focusing of said transducers from on said area, line or point of said

body under examination;

summing said synchronized signals from said transducers;

separating from said summedreceived signals a-components having a second

frequency equal to an even harmonics of said first frequency;

transforming said separated components of said summed signals into image data of

the structure of said body under examination;

displaying said image data by display means; and

wherein said delays are also determined as a function of said even harmonics of

said firstthe frequency-of-said received signals and as a function of the position of said

receiving transducers in said array of receiving transducers.

2.(original) A method as claimed in claim 1, characterized in that the delay

calculation function depends, linearly or non linearly, on the position of each receiving

transducer in the receiving transducer array.

3.(currently amended) A method as claimed in claim 1, characterized in that a term

is added to the delay calculation function for one of said receiving transducers, said term

being determined by the frequency of said secondeven harmonics frequency, said term also

being determined by the position of said one of said receiving transducers in said array of

receiving transducers.

4.(original) A method as claimed in claim 1, characterized in that said delay is a

function of said component of said received signal to be used for image data

transformation, and is such as to cause a phase shift of said received signals, such that

components of said first frequency are suppressed when said synchronized signals are

summed.

5.(currently amended) A method as claimed in claim 1, characterized in that said

first frequency is the fundamental frequency component of said received signals and said

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secondeven harmonic-frequency components of said received signals to be used for

imaging isare the second harmonics frequency of said first frequency.

6.(currently amended) A method as claimed in claim 1, characterized in that said

delays are determined in a manner such that fundamental frequency components of said

received signals are phase shifted by a half-cycle such that said eventhe harmonic

components of saidthe received signals of said receiving transducers are in phase, whereby

said summing of said synchronized signals causes the suppression of said fundamental

frequency components of said first frequency theand said even harmonic components of

said synchronized signals are summed in a non destructive manner to form an amplified

signal.

7.(original) A method as claimed in claim 1, characterized in that said delay for

each of said receiving transducers is determined by using the following function

$$\frac{x_i sen\theta_0}{c} + i \frac{1}{2f_0}$$

where:

"i" = transducer index;

f0:= fundamental frequency;

X<sub>i</sub>:= distance of the transducer "i" from a predetermined reference point;

 $\theta_0$ := steering angle.

8.(currently amended) A method as claimed in claim 6, characterized in that said

summedthe signals resulting from the sum of receive signals is are determined by using the

following equation

$$b(t,\theta_0) = \sum_i s_i \left( t - \frac{x_i \sin \theta_0}{c} - i \frac{1}{2f_0} \right)$$

where: "i" = transducer index;

 $f_0$ := fundamental frequency;

X<sub>i</sub>:= distance of the transducer "i" from a predetermined reference point;

S<sub>i</sub>:= receive signal from the transducer "i";

 $\theta_0$ := steering angle;

 $b(t, \theta_0) := sum signal.$ 

9.(currently amended) A method as claimed in claim 6, characterized in that in

order to at least partly suppress or reduce the non-zero parts of said summed signals, a

change in the phase shift direction of successive received signals is provided with

reference to the moment in which said received signals are received from each of the

receiving transducers of said transducer array, said phase shift being kept substantially

constant for components at said first frequency and for said components at said even

harmonicsecond frequency.

10.(original) A method as claimed in claim 8, characterized in that said function to

calculate said delay for each of said receiving transducers is associated with a phase shift

direction changing sequence in which the elements of said sequence are applied as a function of each of said transducers in said transducer array.

11.(original) A method as claimed in claim 10, characterized in that said function

provides a 0, 1, 0, 1, 0, 1, .... corresponding to a rem(i/2) function, where "i" is the index

of each transducer in the transducer array, in lieu of the simple index "i".

12.(original) A method as claimed in claim 11, characterized in that the function

for calculating said receiver delay for each of said transducers is as follows:

$$\frac{x_i \sin \theta_0}{c} + \operatorname{rem}(i/2) \frac{1}{2f_0}$$

whereas the function for summing the receive signals from all receiving transducers is as follows:

$$b(t,\theta_0) = \sum_{i} s_i \left( t - \frac{x_i \sin \theta_0}{c} - \text{rem}(i/2) \frac{1}{2f_0} \right)$$

where: "i" = transducer index;

 $f_0$ := fundamental frequency;

X<sub>i</sub>:= distance of the transducer "i" from a predetermined reference point;

 $S_i$ := receive signal from the transducer "i";

 $\theta_0$ := steering angle;

b(t,  $\theta_0$ ):= sum signal.

13.(original) A method as claimed in claim 10, characterized in that said function

provides a sequence, 0, 1, 0, 1, 1, 1, .... corresponding to a (rem((i+1)/3)1) function, where

"i" is the index of each transducer in the transducer array.

14.(original) A method as claimed in claim 10, characterized in that said function

provides a sequence including the elements 0, 1, 0, 1, 1, 1, .... and corresponding to a (-

 $1)^{(i+1)/2}$  rem(i/2), where "i" is the index of each transducer in the transducer array.

15.(original) A method as claimed in claim 1, characterized in that said

transmitting comprises at least one pulsed signal having an envelope with smoothed edges.

16.(original) A method as claimed in claim 15, in which said pulsed signal

comprises a triangular envelope.

17.(original) A method as claimed in claim 15, in which said pulsed signal

comprises a Gaussian envelope.

18.(original) A method as claimed in claim 15, characterized in that said envelope

is smoothed by using filters.

19.(original) A method as claimed in claim 1, characterized in that said transmitted

acoustic wave signals are generated within an envelope having smoothed edges.

20.(original) A method as claimed in claim 19, characterized in that said signals

comprise a triangular envelope.

21.(original) A method as claimed in claim 19, characterized in that said signals

comprise a Gaussian envelope.

22.(original) A method as claimed in claim 1, characterized in that said delays are

also calculated as a function of the distance of said reflection sources from the origin of a

selected coordinate system which describes the ultrasonic beam propagation space.

23.(currently amended) A method as claimed in claim 1, characterized in that in

the calculation of said delays includes a term relating to the value of said first frequency-f<sub>0</sub>

which is chosen to be greater than the effective-value of said first frequency, said first

frequency being the fundamental frequency of said transmitted beams.

24.(currently amended) A method according to claim 23, characterized in that the

value of said term- $f_0$  is increased by between 25% to 50% of the effective value of said

firstfundamental frequency of said transmitted beams.

25.(currently amended) A method according to claims 23, characterized in that

high-pass filtering of said summed signal is carried out with a cutting frequency lying

between said fundamental frequency and said even harmonicsecond frequency, said even

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harmonic second frequency being the frequency of the second harmonic of said transmitted

beams.

26.(original) An ultrasonic imaging apparatus comprising:

at least one ultrasonic probe having a plurality of transmitting transducers for

generating transmission beams, and a plurality of receiving transducers;

a beamformer coupled to said receiving transducers for applying receiver signal

synchronization delays to each of said receiving transducers, said delays being determined

with reference to the direction in which said transducers are focused;

means for processing received signals from each of said receiving transducers,

including means for attenuating the fundamental frequency component of said received

signals;

means for summing said received signals from their respective ones of said

receiving transducers;

means for transforming said summed signals into image data;

display means for displaying said image data in the form of graphic images; and

a programmable control means for controlling said beamformer, said control means

comprising one or more algorithms used for calculating said delays, said delays being

calculated as a function of the position of said transducer in said transducer array, said

delays further being calculated with respect to a predetermined reference point, based on

the steering angle, on the focusing distance and on a predetermined harmonic of the

fundamental frequency of said received signals.

27.(currently amended) An apparatus as claimed in claim 26, characterized in that

said beamformer is programmed or controlled by said control means that is programmable

to calculate said delays for each of said receiving transducers in order to generate a change

in the phase of said received signals.

28.(currently amended) An apparatus as claimed in claim 27, characterized in that

said phase change is caused by the functional dependence of delays from the selected said

predetermined harmonic frequency, particularlysaid predetermined harmonic frequency

being the second harmonic frequency of said received signals.

29.(original) An apparatus as claimed in claim 26, characterized in that said

beamformer is programmed or controlled by said control means in such a manner as to

calculate received signals for each of said receiving transducer according to the following

function:

$$\frac{x_i \sec \theta_0}{c} + i \frac{1}{2f_0}$$

where: "i" = transducer index;

 $f_0$ := fundamental frequency;

 $X_i$ := distance of the transducer "i" from a predetermined reference point;

 $\theta_0$ := steering angle.

30.(currently amended) An apparatus as claimed in claim 26, characterized in that

said beamformer is controlled by said control means in such a manner as to combine the

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phase shift of said received signals from each of said transducers, said phase shift being

caused by the application of functional delays and by the dependence thereof from the

selected said second harmonic frequency, particularly the second harmonic, with a phase

shift direction changing sequence, composed of alternate "0" and "1" elements.

31.(original) An apparatus as claimed in claim 30, characterized in that said phase

shift changing sequence is defined by a rem (i/2) function, where "i" is the index of the

transducer in the transducer array, the delay calculation algorithm being as follows:

$$\frac{x_i \sec \theta_0}{c} + \operatorname{rem}(i/2) \frac{1}{2f_0}$$

where: "i" = transducer index;

 $f_0$ := fundamental frequency;

 $X_i$ := distance of the transducer "i" from a predetermined reference point;

 $\theta_0$ := steering angle.

32.(original) An apparatus as claimed in claim 30, characterized in that said phase

shift changing sequence is  $0, 1, 1, 0, 1, 1, \dots$  corresponding to a (rem((i+1)/3)1) function,

where "i" is the index of each transducer in the transducer array.

33.(original) An apparatus as claimed in claim 30, characterized in that said phase

shift changing sequence includes the elements 0, 1, 0, 1, 0, 1, 0, 1, .... corresponding to a

(-1)<sup>(i+1)/2</sup>rem(i/2) function, where "i" is the index of each transducer in the transducer array.

34.(original) An apparatus as claimed in claim 26, further comprising means for

generating in at least one of said transmission beams a pulse comprising an envelope with

smoothed edges.

35.(original) An apparatus as claimed in claim 34, wherein said pulse comprises a

triangular envelope.

36.(original) An apparatus as claimed in claim 34, wherein said pulse comprises a

Gaussian envelope.

37.(original) An apparatus as claimed in claim 26, further comprising means for

generating a smoothed envelope in the echoes contained in one or more or said received

signals.

38.(original) An apparatus as claimed in claim 37, wherein said echoes form a

triangular envelope.

39.(original) An apparatus as claimed in claim 37, wherein said echoes form a

Gaussian envelope.

40.(original) An apparatus as claimed in claim 26, further comprising means for

increasing the value of the fundamental frequency  $f_0$  in the delay calculation equations.

41.(original) An apparatus according to claim 40, characterized in that the value of

the term f<sub>0</sub> is increased to between 25% to 50% of the effective value of the fundamental

frequency of said transmission beams.

42.(original) An apparatus according to claim 26, characterized in said summing

means comprises a high-pass filter having a cutting frequency lying between the

fundamental frequency and the frequency of the second harmonic of said received signals.